EMF Exposure Limits and Compliance Assessment for Wireless Devices Operating at Frequencies above 6 GHz

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Work Package 2

- the design of a novel miniature field probe for assessing the near E-field of antennas in portable devices used in close proximity of the body, operating in the band 10-100 GHz
- addressing:
 - test equipment
 - spatial averaging
 - sampling resolution

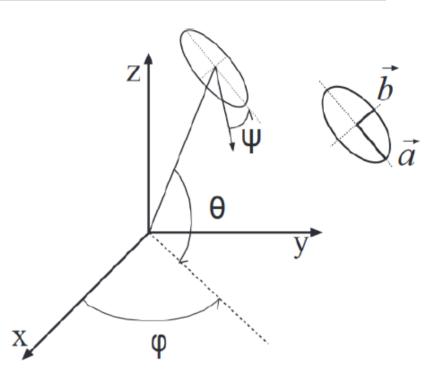
Near-Field Probe System

- pseudo-vector probe design
- designed for optimised isotropy
- insensitive to mechanical tolerances and field distortion
- probe developed by SPEAG
 - commercial availability: end of Q3 2016
 - open specification for measuring systems
- acquired by IT'IS for this project



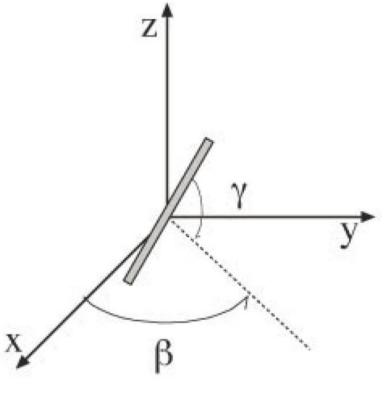
Pseudo Vector Probes

- reconstructs the polarization ellipse by oversampling the field
- at least 5 measurements are required
- the proposed analysis uses the E-field magnitude and the normal to the polarization ellipse defined in terms of θ and φ.



Pseudo Vector Probes

- vector magnitudes a and b can be uniquely determined from least-square error for the given set of angles
- overdetermined system of equations to suppress noise and increase the reconstruction accuracy
- solution:
 - probe with two sensors
 - angles: γ_1 , $\gamma_2 = \gamma_1 + 90^\circ$
 - perform measurements at three angular rotations: β_1 , β_2 , β_3
 - angles optimized for polarization accuracy



Pseudo Vector Probe: Method

• parameters a, b, ψ , ϕ , θ are evaluated from

$$f_{ij}^2 = a^2 \cdot [\chi_1 - \chi_2 - \chi_3]^2 + b^2 \cdot [\chi_4 + \chi_5 - \chi_6]^2$$

where

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$$\chi_{1} = \cos(\psi) \cdot \cos(\theta) \cdot \sin(\gamma_{i})$$

$$\chi_{2} = \cos(\gamma_{i}) \cdot \sin(\psi) \cdot \sin(\beta_{j} - \phi)$$

$$\chi_{3} = \cos(\gamma_{i}) \cdot \cos(\psi) \cdot \sin(\theta) \cdot \cos(\beta_{j} - \phi)$$

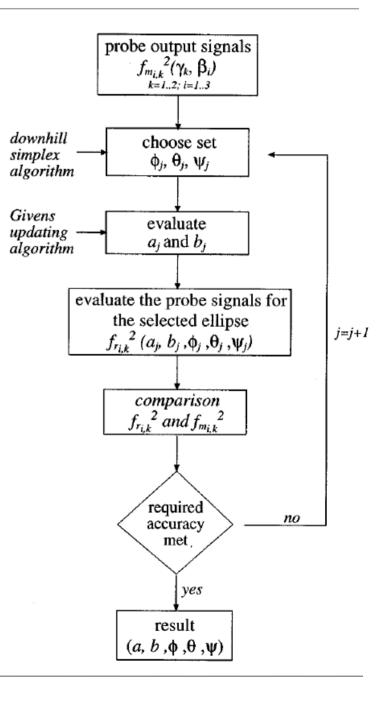
$$\chi_{4} = \sin(\psi) \cdot \cos(\theta) \cdot \cos(\gamma_{i})$$

$$\chi_{5} = \cos(\gamma_{i}) \cdot \cos(\psi) \cdot \sin(\beta_{j} - \phi)$$

$$\chi_{6} = \cos(\gamma_{i}) \cdot \sin(\psi) \cdot \sin(\theta) \cdot \cos(\beta_{j} - \phi)$$

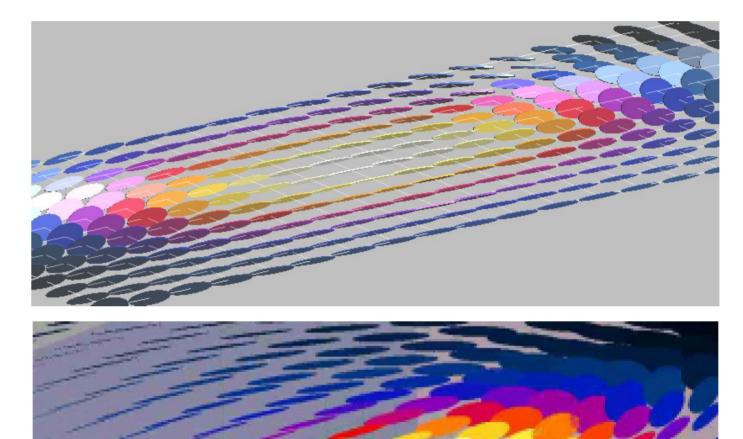
 apply Givens algorithm [1] with downhill simplex optimization

[1] G. H. Golub and C. F. Van Loan, *Matric Computations*. Baltimore, MD: Johns Hopkins Univ. Press, 1990.



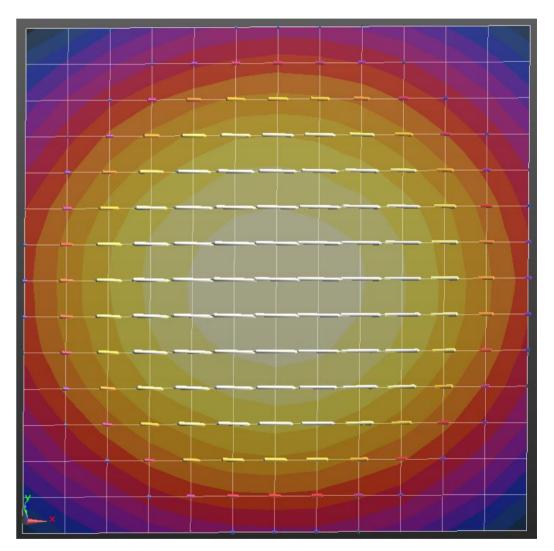
Validation of Concept

- amplitude and polarization of E field
- measurement (top) vs simulation (bottom)



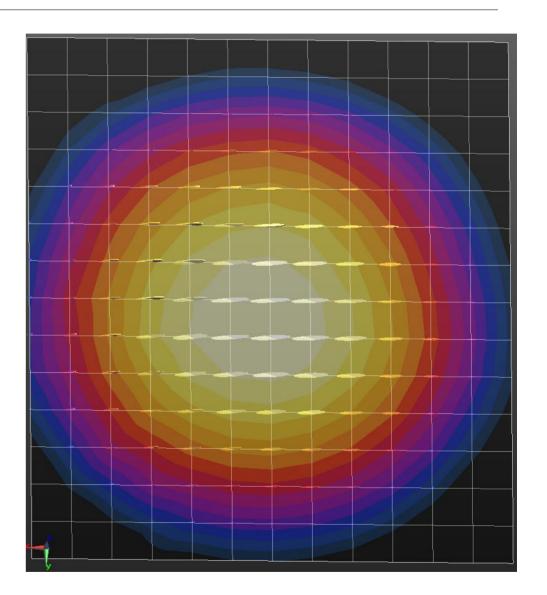
Pseudo-Vector Field Scans (6.5 – 18 GHz)

- field scan 10 cm above 10 GHz horn
- scan 60 x 70 mm step 5 mm
- field distribution verified



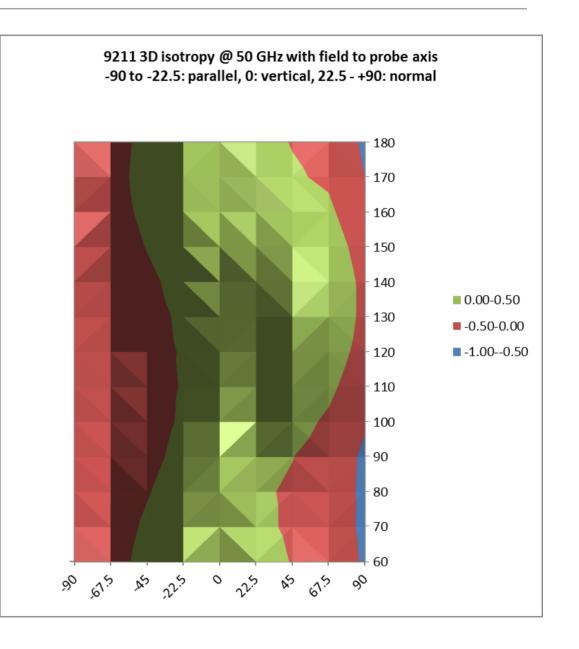
Pseudo-Vector Field Scans 50 - 75 GHz

- field scan 15 cm above 50 GHz horn
- scan 60 x 70 mm step 5 mm
- field distribution verified



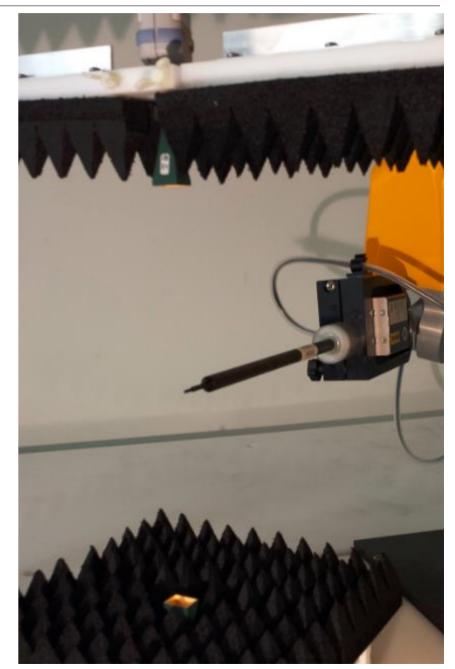
EUmmW Parameters

- example for 3D isotropy
 - probe at 50 GHz
 - all orientations above horn
 - plot for total field
- spherical isotropy deviation < 0.5 dB
 - < 0.2 dB below 30 GHz</p>
- dynamic range
 - 30 V/m to > 1000 V/m



Calibration Process

- three antenna method
 - 2 horn antennas for transmitter and receiver
 - probe as third antenna
 - advantages over TEM cell or waveguide methods
- two-step process
- step 1:
 - align transmit & receive horns
 - characterize electric field at fixed point
 - measure at 3 different heights
 - determine phase center vs frequency
- step 2:
 - remove receive horn
 - insert probe at calibration point
 - probe is outside reactive near field



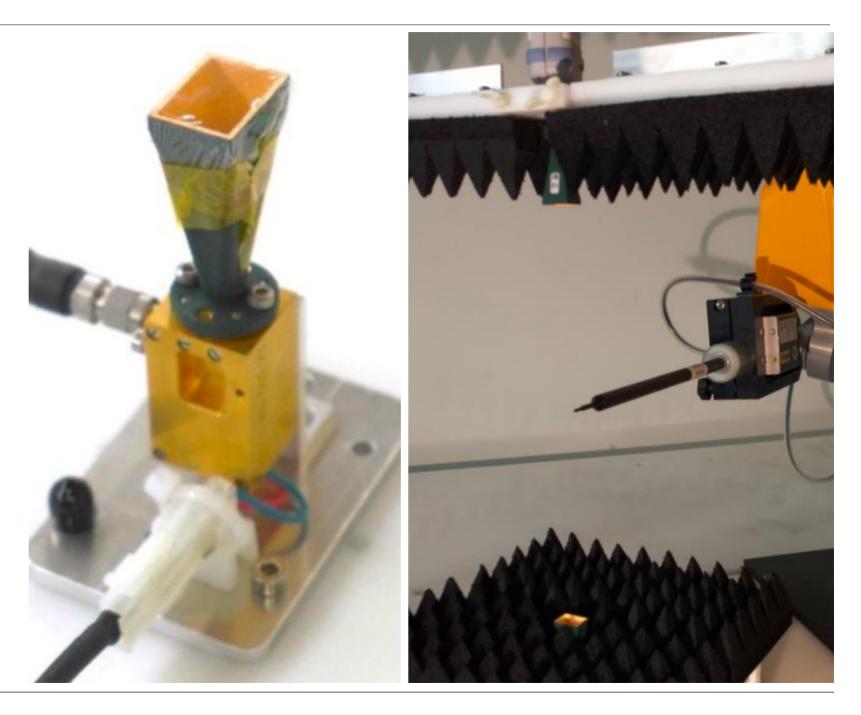
Calibration Method: 6.5 – 18 GHz

- 2 identical horns
 - specified gain
- 3 fixed horn distances
 - field calibration
- calibrated input power
 - field verification
- monitoring antenna
 - removable
 - reduces reflections
- full 3D probe movement



Calibration Method: 50 – 75 GHz

 identical concept as lower band



Conclusion – Calibration System

- calibration method developed
- calibration uncertainty: < ± 1.0 dB
- frequency range: 6 75 GHz
- ISO/IEC 17025 accreditation
 - accredited for frequencies below 6 GHz
 - accreditation in process for frequencies above 6 GHz
- next step: publication of method and calibration

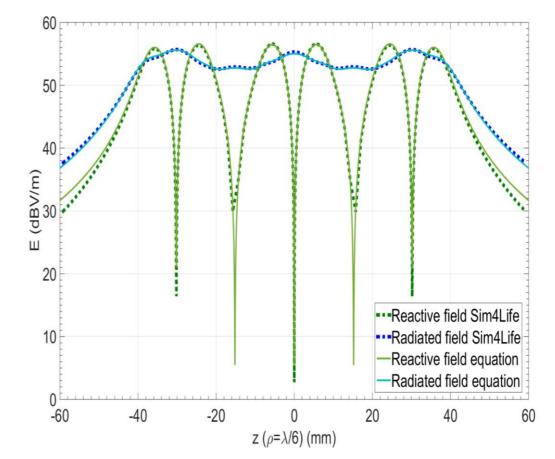
Near Field of Arrays: Unexplored Topic

- new theory of electromagnetic fields near dipole arrays developed and tested by extensive numerical computations.
 - also applicable to slot arrays
- reactive near fields (RNF) extend only as far as RNF of individual array elements. Distance (≈λ/2π) much smaller than previously predicted. In RNF zone, non-propagating reactive e-fields are predominant.
- at λ/2 distance, radiated E-fields are predominant. Radiative near fields extend out to antenna-sized dependent Fraunhofer zone (far field).
- near E-fields' nature dictates type of power density measurements near arrays.

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Dosimetry Needed in Reactive Near Field

- regions defined by ratio: E_z / E_ρ
- formulas derived for electric dipoles
- reactive near field: $d < \lambda/6$
 - $E_z \ll E_\rho$
 - power density measurements not indicated
 - reactive coupling between exposed body and antenna alters free field distribution of current and charges on radiating elements
 - dosimetric measurement needed as at lower frequency bands
- distance much lower than estimated by Balanis (based on phase change)

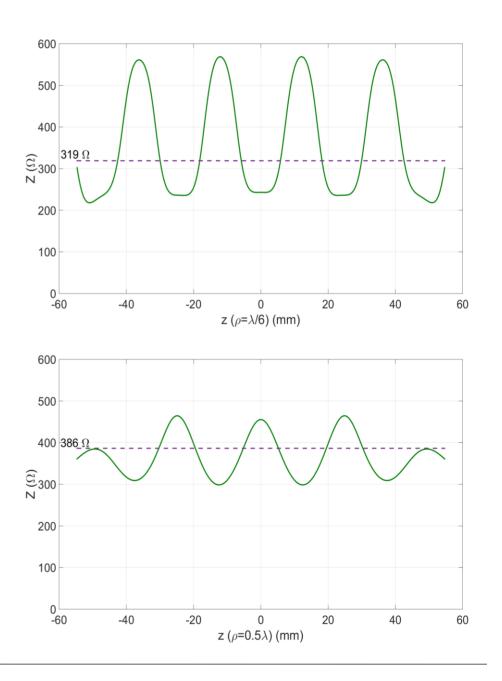


Power Density in Radiative Near Field & Far Field

- transition zone: $\lambda/6 < d < \lambda/2$
 - field polarization needed
 - pseudo-vector electric field probes can be used
- radiative near field: $d > \lambda/2$
 - $E_z >> E_\rho$
 - power density measurements
 - scalar electric field probes or waveguides may be used
- 50 40 (dBV/m) 00 ш 20 Reactive field Sim4Life Radiated field Sim4Life 10 Reactive field equation Radiated field equation -40 -20 -60 20 40 60 $z(\rho=\lambda/2)$ (mm)
- resolve field polarization to determine upper bound on exposure metric

Impedance in Radiative Near Field & Far Field

- impedance calculated
 - E_z / H_y
- 5-element collinear dipole array
 - 10 GHz
 - 0.8 λ separation
- two distances
 - d = $\lambda/6$: Z variation > 3 dB
 - d = $\lambda/2$: Z variation < 2 dB
- λ/2 seems to be minimum distance for measurements with scalar E-field probes or waveguides
 - e.g., 15 mm @ 10 GHz, 2.5 mm @ 60 GHz



Proposed Procedure

- measurement of the field at the closest distance from the device
 - pseudo-vector near field with EUmmW probe
 - system integration: cDASY6 or DASY52 as test bed
 - adaptive scanning resolution from λ / 10 (depending on distribution)
- determination of the radiative & reactive components
- evaluation of the field with respect to human exposure and regulatory requirements
 - transformation matrix to plane-wave equivalence
 - includes appropriate spatial averaging
- determination compliance and generation of report

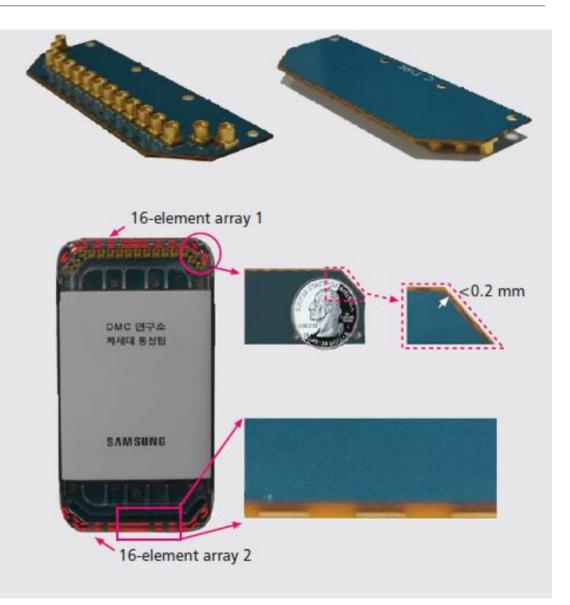
Comprehensive Validation

- comprehensive set of sources with respect to
 - frequency
 - field distribution

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Implementations of Millimeter Wave Antennas

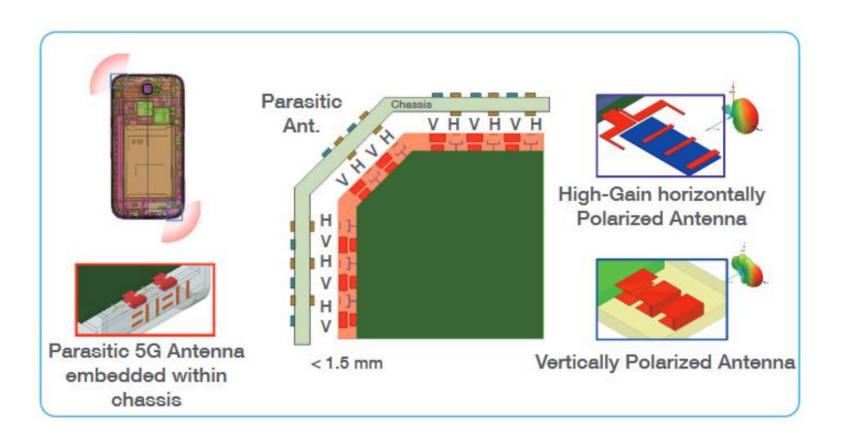
- array of rectangular patches at 60 GHz
- 16 elements
- patch width: 5 mm (λ / 2)
- placed on top or bottom of device



T. S. Rappaport, W. Roh, K. Cheun, "Mobile's millimeter-wave makeover," IEEE Spectrum, Vol. 51, nol. 9, pp. 34-58, Sept. 2014.

Implementations of Millimeter Wave Antennas

- similar design as previous case
- parasitic elements integrated into the phone chassis



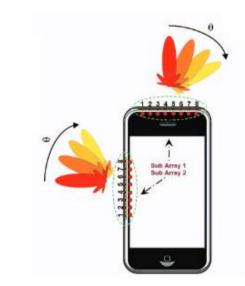
SAMSUNG 5G Vision White Paper, Feb. 2015.

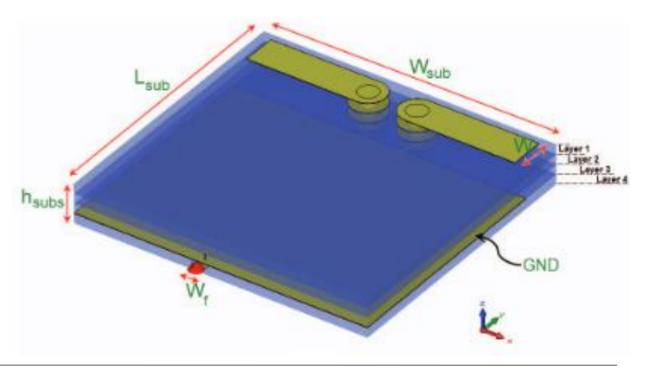


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Implementations of Millimeter Wave Antennas

- multi-layered phased array
- off-center dipole antenna elements for MIMO 5G
- frequency: 27 29 GHz.



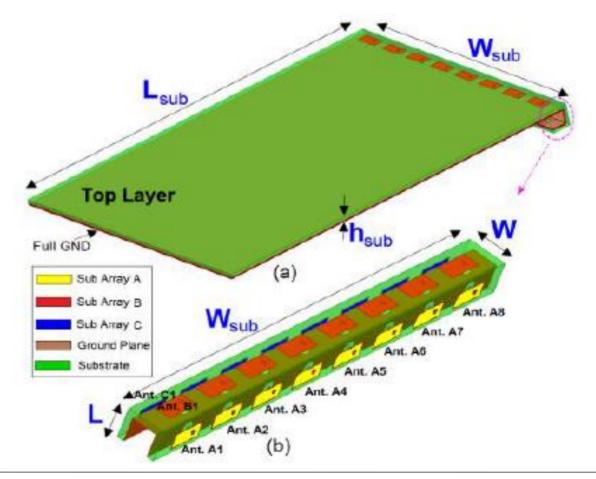


N. Ojaroudiparchin, M. Shen, G. F. Pedersen, "Multi-Layer 5G Mobile Phone Antenna for Multi-User MIMO Communications," Proc. 23rd Telecommunications Forum (TELFOR), Nov. 2015.

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Implementations of Millimeter Wave Antennas

- 3D-coverage phase array for 5G
- frequency: 21.5 GHz
- uses 3 sub arrays of microstrip patches (8 patches for each sub array)
- half wavelength spacing
- faces at 0°, 90° and 180°



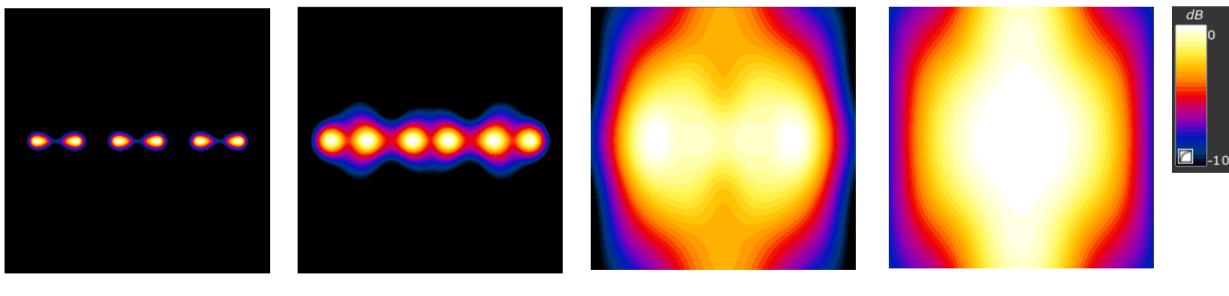
N. Ojaroudiparchin, M. Shen, S. Zhang, G. F. Pedersen, "A Switchable 3D-Coverage Phased Array Antenna Package for 5G Mobile Terminals," IEEE Antennas and Wireless Propag. Lett., 2016.

Millimeter Wave Sources Investigated

- Linear array of collinear half-wavelength dipoles (3 elements)
- Linear broadside array of half-wavelength dipoles (3 elements)
- Linear array of collinear half-wavelength slots (3 elements)
- Linear array of half-wavelength square patches (3 elements)
- Planar array of half wavelength patches (9 elements)
- frequencies: 10, 24, 60, 90 GHz

Sources Used for Validation of Method

- 3-element array of collinear $\lambda/2$ dipoles spaced 0.8 λ
- E_{RMS} field 10 GHz



Plane z=-0.03λ (-1.0 mm)

Plane $z=-0.1\lambda$ (3.0 mm)

Plane z=-0.5 λ (-15 mm)

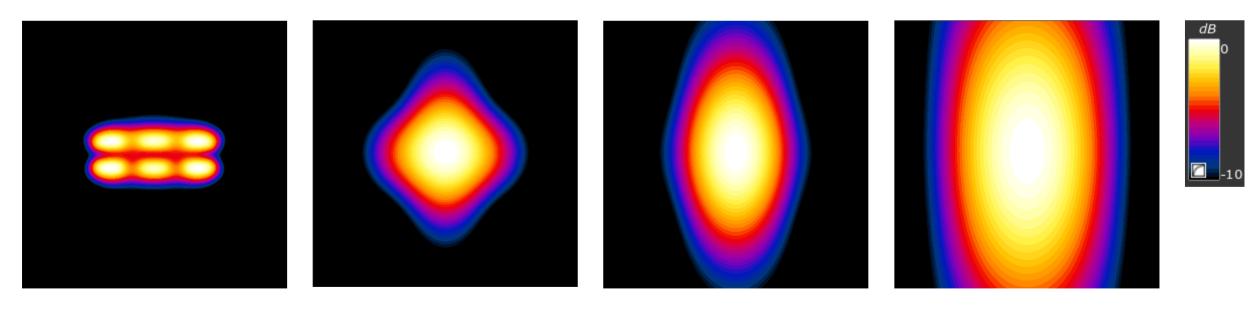
Plane z=-1.0λ (-30 mm)

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Sources Used for Validation of Method

- 3-element array of half-wavelength patches,
- + E_{RMS} field 60 GHz





Plane z=-0.1 λ (-0.5 mm)

Plane z=-0.5λ (-2.5 mm)

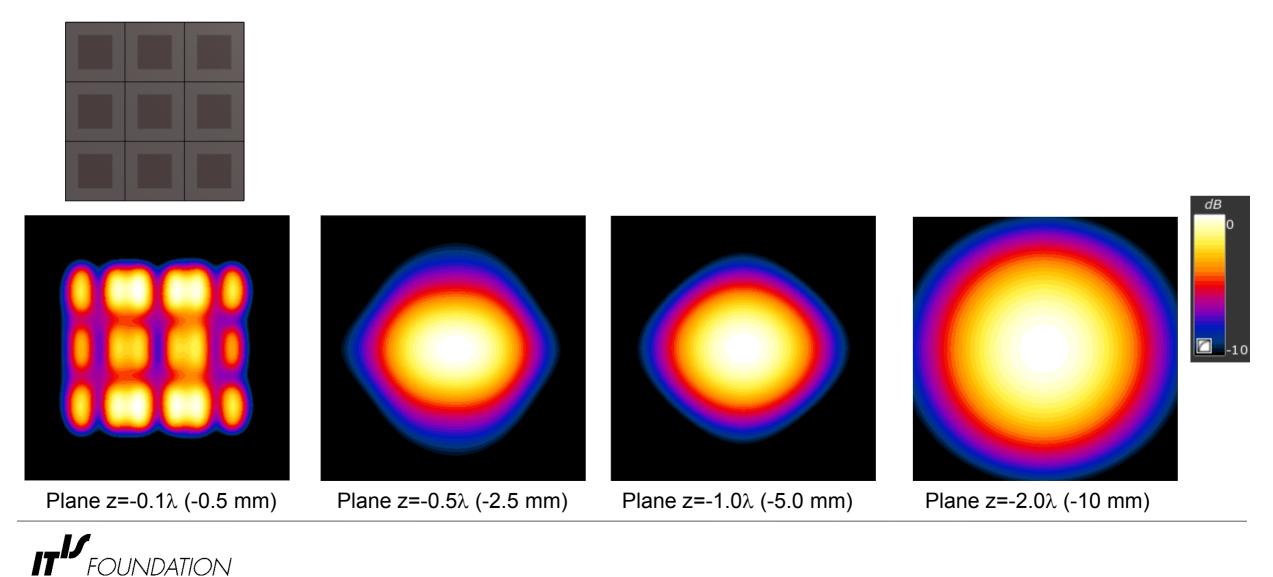
Plane z=-1.0λ (-5.0 mm)

Plane z=-2.0λ (-10 mm)

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Sources Used for Validation of Method

- 9-element array of half-wavelength patches
- ERMS field 60 GHz



Publications (Current and Planned) - WP2

- Q. Balzano, K. Foster, M. Ziskin, D. Colombi, E. Carrasco, M. Capstick, "Dosimetry Considerations for Wireless Devices Above 6 GHz," in progress, planned for IEEE Trans. Vehicular Technology, Q3 2016.
- S. Kühn, M. Douglas, N. Kuster, "Compliance in the Near Field at Millimeter Wave Frequencies using Pseudo-Vector Probes," in progress, Q3 2016.

Conclusions: WP2

- evaluation of measurement hardware
 - pseudo-vector probe developed from 100 MHz to 75 GHz
 - preliminary integration into DASY52 as test bed
- calibration system
 - 3-antenna system (6 75 GHz) developed
 - preliminary uncertainty budget completed
 - ISO/IEC 17025 accreditation extension submitted
- near field to power density transformation
 - new theory developed for determination of near / far field regions
 - development of plane-wave equivalent transformation matrix in progress
- next steps
 - complete transformation matrix (plane-wave equivalence)
 - measurement protocol with compliance criteria implemented in DASY52 and cDASY6
 - publication