

**PROPEL 5G** 

**PRO**cedures and techniques for Proving compliance to Exposure Limits of **5G** wireless devices

A Mobile Manufacturers Forum sponsored project

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- → Response to the MMF RfP on EMF exposure limits and compliance assessment for wireless devices operating at frequencies above 6 GHz – WP2: Compliance testing at frequencies above 6 GHz
- → Focus on accurate evaluation of free-space near-field power density (PD)
  - No discussion on phantoms
  - No discussion on the relevance of basic restrictions on PD
- → Development of concepts generally relevant to frequencies above 6 GHz but mainly validated and applied in the WiGig frequency range (57 – 66 GHz)



- → Core objective: evaluate concepts, techniques and procedures for assessing the near-field power density in air in the very nearfield of wireless devices
- → Create accurate numerical models of existing 60 GHz antennas to be used as near-field references (SO2.1)
- → Assess the accuracy of near-field reconstruction algorithms and determine optimal algorithm parameters (SO2.2)
- → Build up a planar scanner and apply the most relevant identified techniques to the evaluation of near-field PD for reference devices (SO2.3)

# ART-FI

#### $\rightarrow$ Probing distance

- A large distance may not allow efficient reconstruction of the very near-field: system sensitivity, loss of reactive field information
- A short probing distance would induce unacceptable perturbations of the DUT
- → Measurement sampling rate
- → Truncation
- $\rightarrow$  PD averaging area
- → Probe correction
- $\rightarrow$  Usability of phased methods and sensitivity to phase measurement error
- $\rightarrow$  Comparison between phased and phaseless approaches

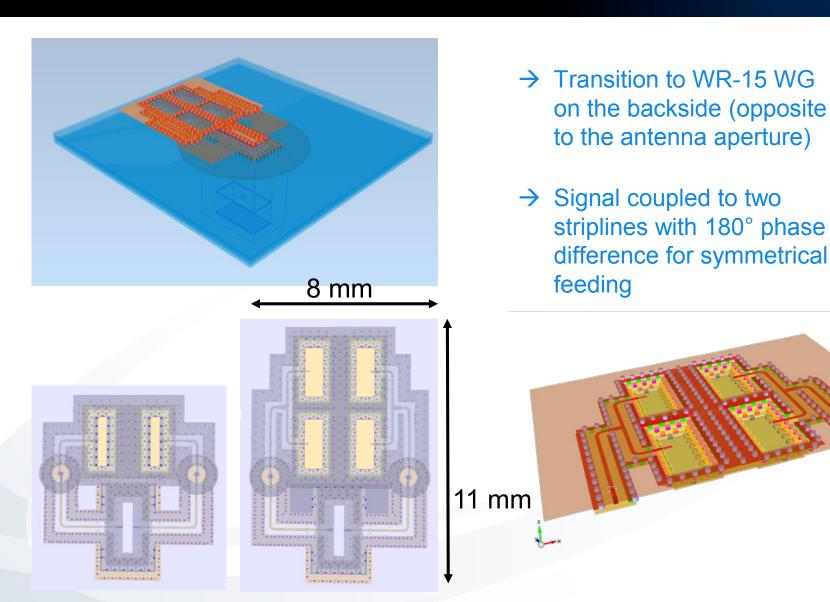


- → Two reference antennas available for both simulations and measurements used for testing near-field algorithms and measurement methods
  - 60 GHz 2x2 LTCC open-ended waveguide antenna array
  - 60 GHz Tx / Rx double-patch antenna
- → Simulations carried out with IEC 62704 compliant Finite Difference Time Domain (FDTD) software EMPIRE by IMST GmbH



#### > LTCC Array Antenna Structure Overview





# > LTCC Array Antenna Sample



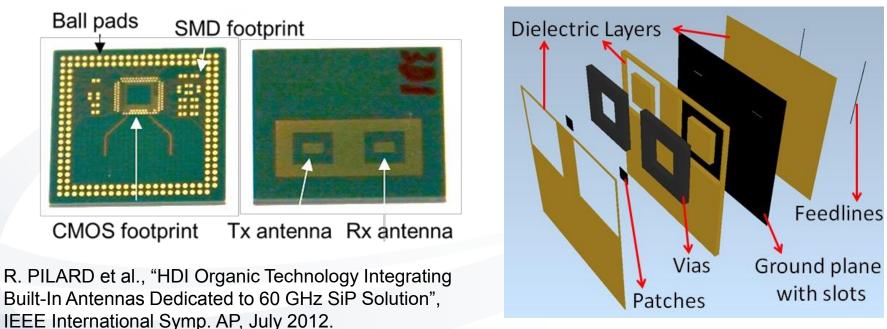


# > Tx/Rx Patch Antenna Array



- → Total Size: 22 mm x 26 mm x 16 mm
- → Chip Size: 12 mm x 12 mm
- → Operating frequency: 57-66 GHz

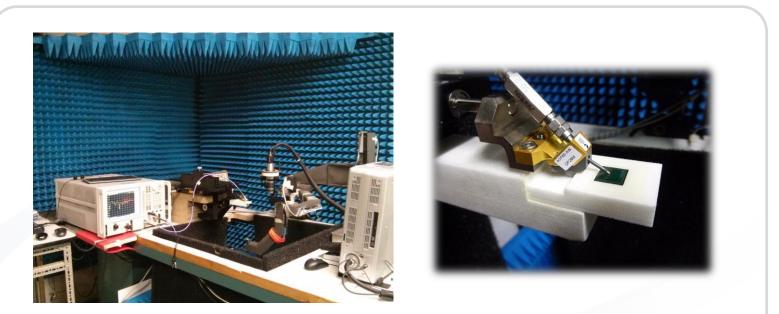








→ Dedicated probe-fed antenna setup for S11 and 3D radiation pattern up to 140 GHz.



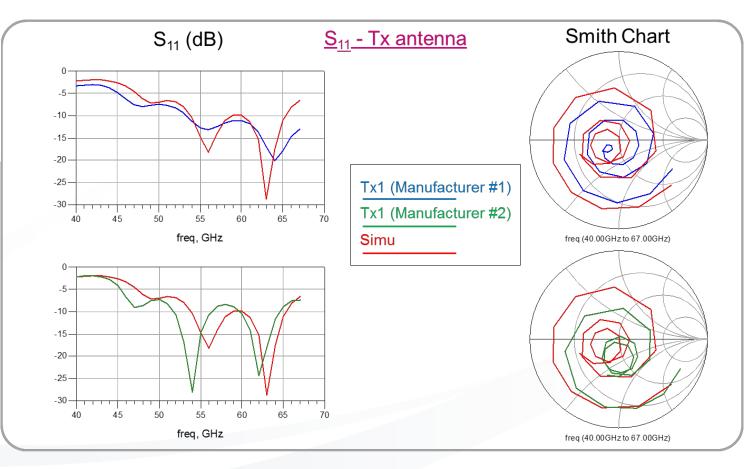
D. Titz et al, "Development of a Millimeter-Wave Measurement Setup and Dedicated Techniques to Characterize the Matching and Radiation Performance of Probe-Fed Antennas", *IEEE Antennas and Propagation Magazine*, vol. 54, pp. 188-203, 2012.

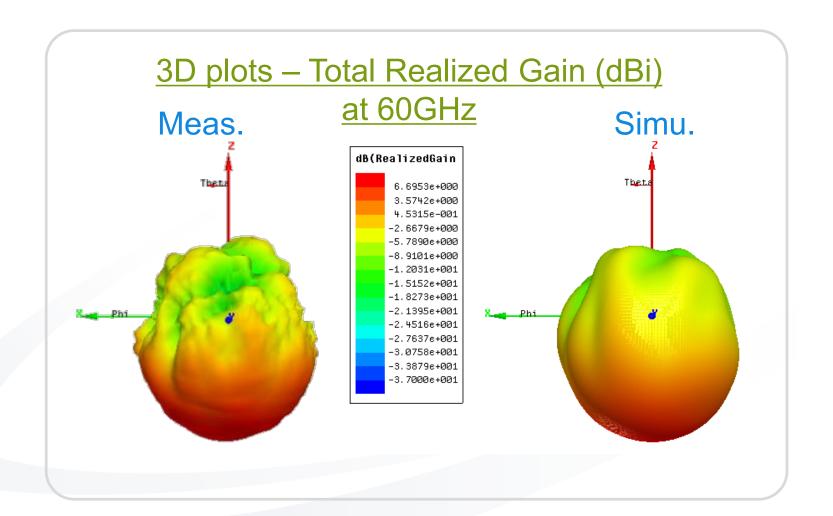
A. Bisognin et al.," Probe-fed measurement system for F-band antennas", EuCAP 2014.

#### Tx/Rx Patch Antenna Array Measurements vs Simulations 1/2









ART-F



- → The time-integrated value of Poynting vector S(t) represents the energy flow of the electromagnetic field during the considered period of time at the considered location
- → Near-field PD in harmonic regime can be expressed as the real part of the complex Poynting vector

$$\underline{S}(t) = \underline{E}(t) \times \underline{H}(t)$$

$$\underline{E}(t) = \underline{E}_{\underline{m}} e^{j\omega t} ; \underline{H}(t) = \underline{H}_{\underline{m}} e^{j\omega t}$$

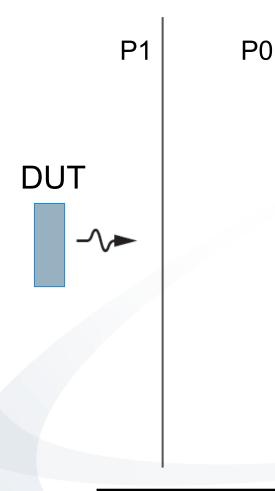
$$\left\langle \underline{S}(t) \right\rangle = \frac{1}{T} \int_{0}^{T} \underline{S}(t) dt = \operatorname{Re}\left(\frac{1}{2} E_{\underline{m}} \times H_{\underline{m}}^{*}\right)$$

→ The near-field PD can be looked at locally or averaged over a surface through which the power is flowing (ICNIRP – 1 cm<sup>2</sup>)



- → Two components of the vector E or H field (magnitude and phase) over a surface are enough to reconstruct the 6 components of the E and H field over the volume enclosed within the surface
- → The third component of E or H can be calculated from the two other components using Gauss law ( $\nabla$ . <u>E</u> =0 in a source-free volume)
- → H can be calculated from E using Maxwell-Faraday equation:  $\nabla \times \underline{E} = \mu \frac{d}{dt} \underline{H}$
- → The field over the volume is determined from the field over a surface by Huygens principle / surface equivalence theorem





- → E-field measurement taken at plane P0, two components tangential to plane
- → Tangential E-field back-propagated to P1 using plane-wave expansion
- → Normal E-field component reconstructed using Gauss law
- → H-field reconstructed from E using Maxwell-Faraday
- $\rightarrow$  PD obtained from E and H vector fields



#### $\rightarrow$ Tests on Tx / Rx patch antenna array

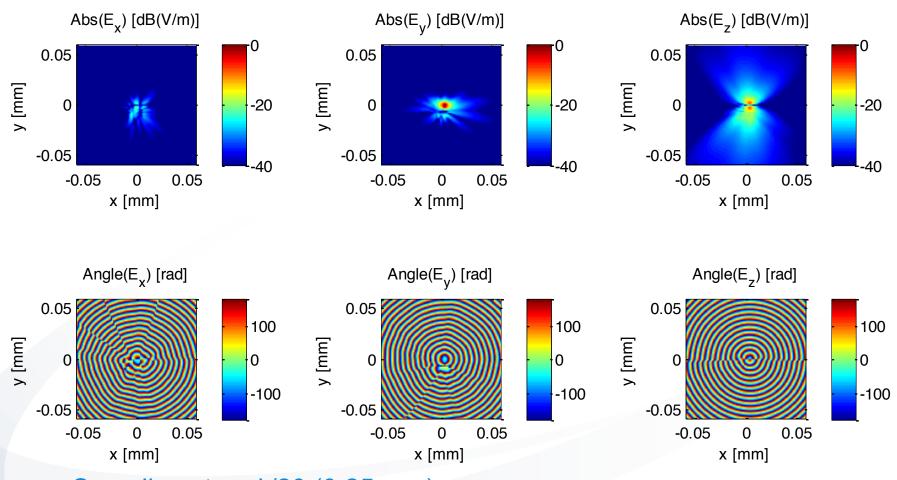
- Frequency: 60 GHz
- **Exploited quantities:** complex 3D distributions of the DUT radiated E-field and H-field computed with EMPIRE FDTD code
- Spatial range of the 3D field distributions: 118x118x30mm<sup>3</sup>

#### → Test description

- A 2D-scan is extracted from the simulated E-field and the H-field, with a fixed sampling rate in the x and y directions (tangential)
- The PWS of E-/H-field is computed for back-propagation till a target plane at a distance h close to the DUT
- The reconstructed and FDTD simulated E, H fields and power density at the distance *h* are compared

## FDTD Simulated E-Field Data at 0 mm from DUT

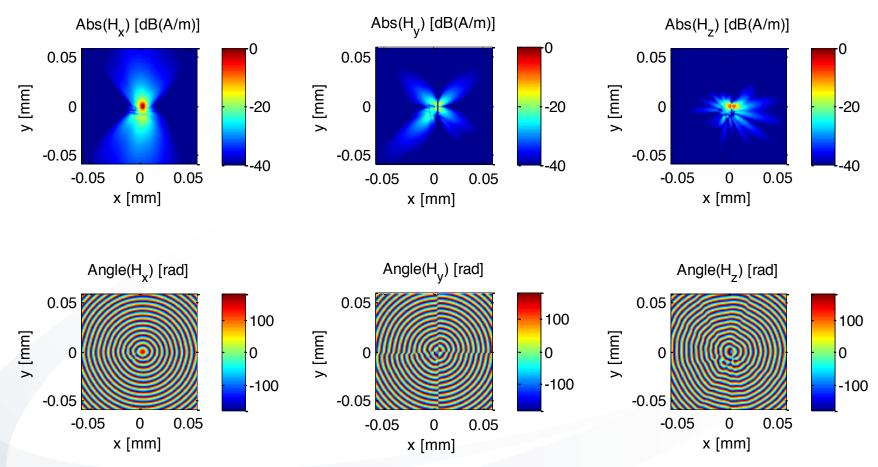




## Sampling step: $\lambda/20$ (0.25 mm) Magnitude normalized to max E-field

## FDTD Simulated H-Field Data at 0 mm from DUT

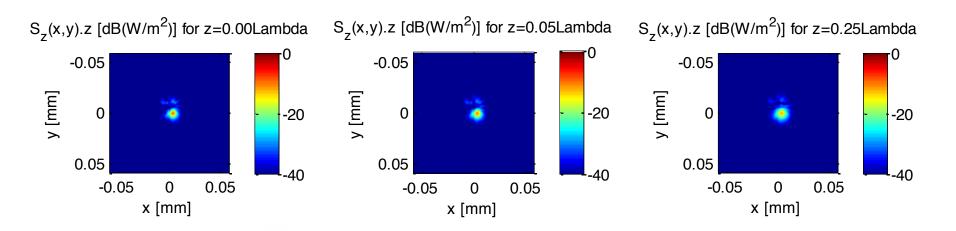


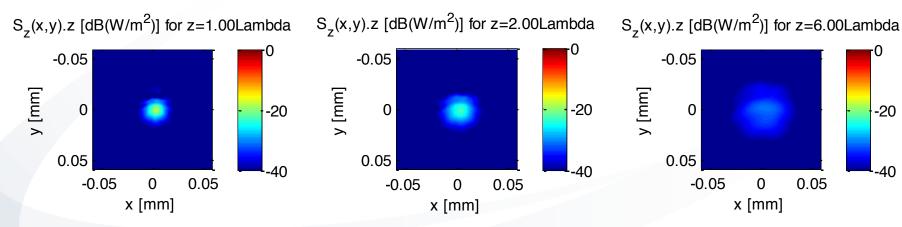


### Sampling step: $\lambda/20$ (0.25 mm) Magnitude normalized to max E-field

# FDTD Simulated PD Through xy Plane







#### Sampling step: $\lambda/20$ (0.25 mm) Normalized to max S magnitude

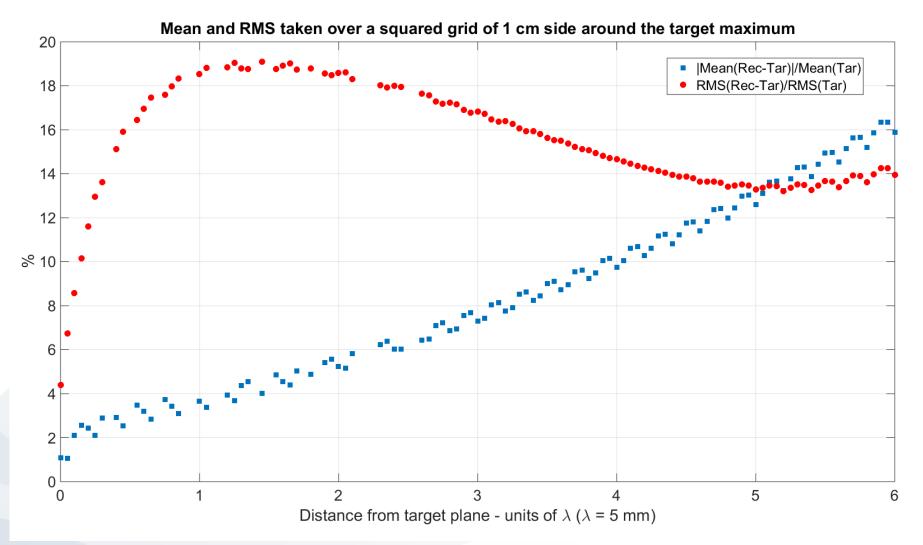
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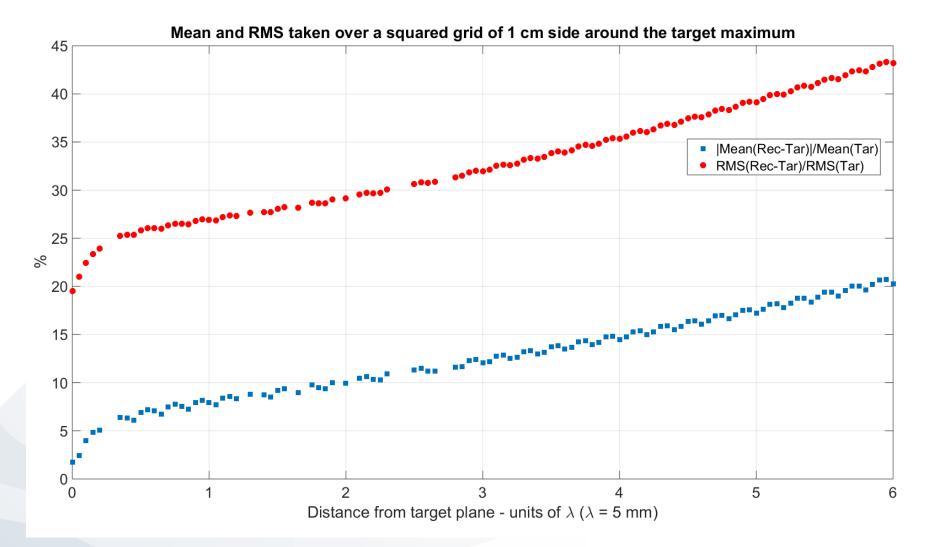
### Deviation of PD Reconstruction at 2 mm from the DUT - $\lambda/20$ Sampling Rate





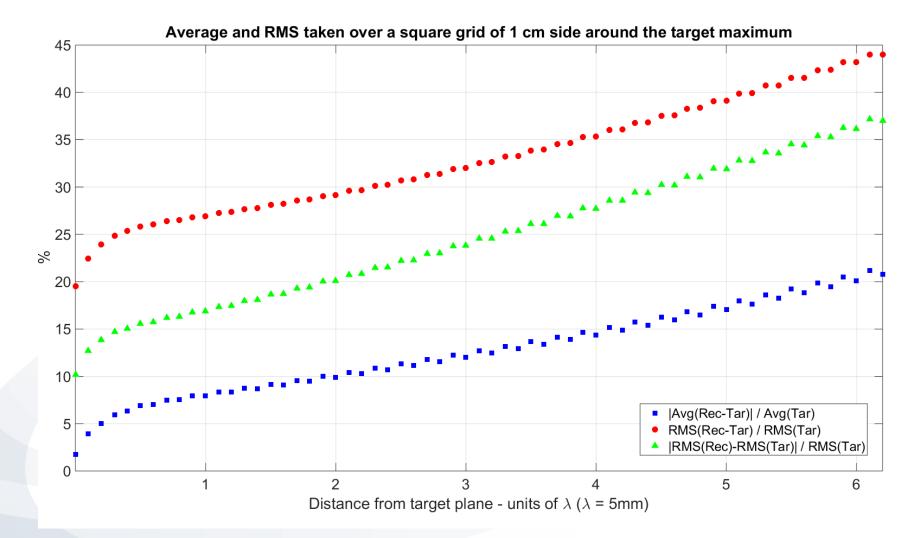
### Deviation of PD Reconstruction at 1 mm from the DUT - $\lambda/20$ Sampling Rate





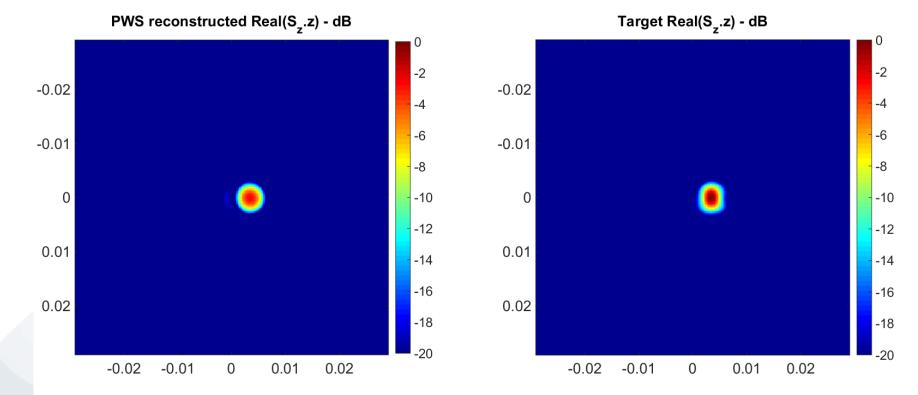
### Deviation of PD Reconstruction at 1 mm from the DUT - $\lambda$ /10 Sampling Rate





#### Reconstructed PD with Measurement Plane Taken at 20 mm from Target Plane





MMF Workshop on Exposure from Future Wireless Devices above 6GHz, BioEM Conference, Ghent Belgium

## Comparison between Simulated and Reconstructed H-field at 7mm from DUT

0.05

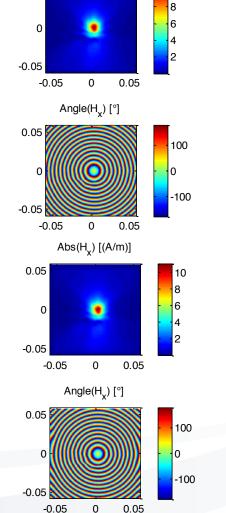
Abs(H<sub>v</sub>) [(A/m)]

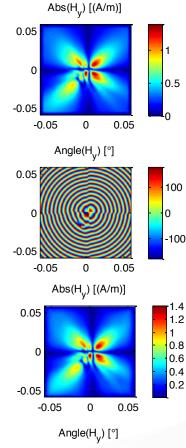
10

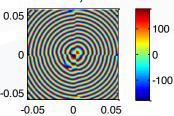


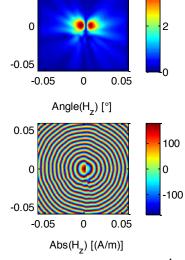
# Simulation

Reconstruction



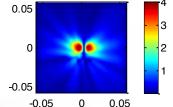


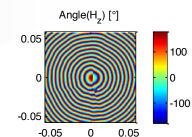




 $Abs(H_{7})[(A/m)]$ 

0.05

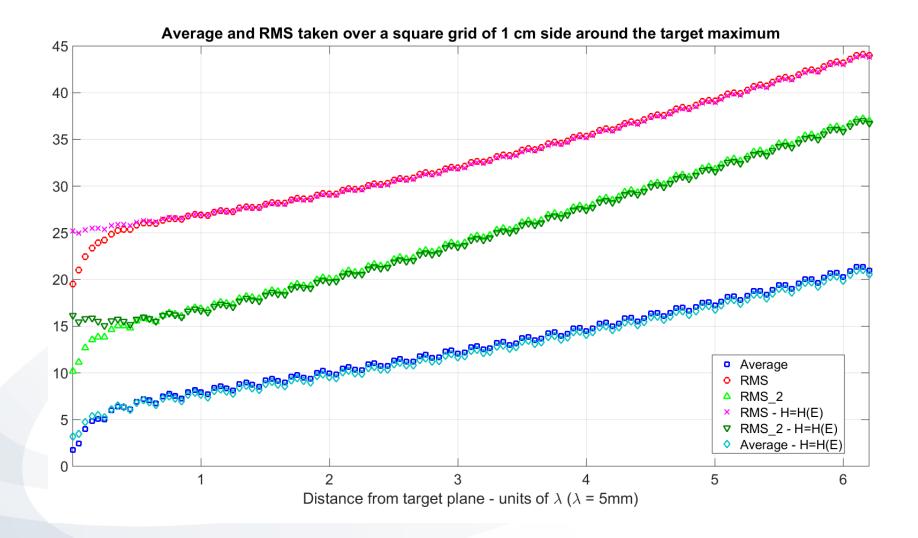




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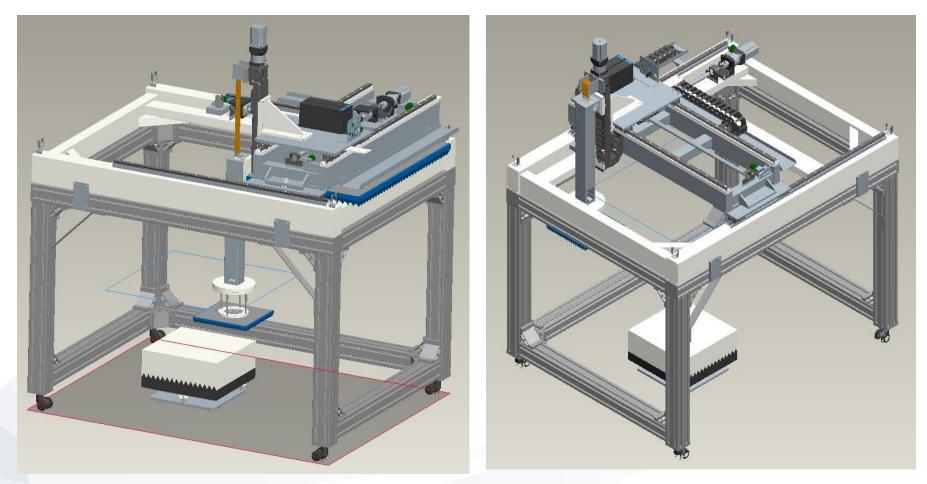
## Impact of H Reconstruction on PD Reconstruction Accuracy





## > NF PD Measurement Setup Construction

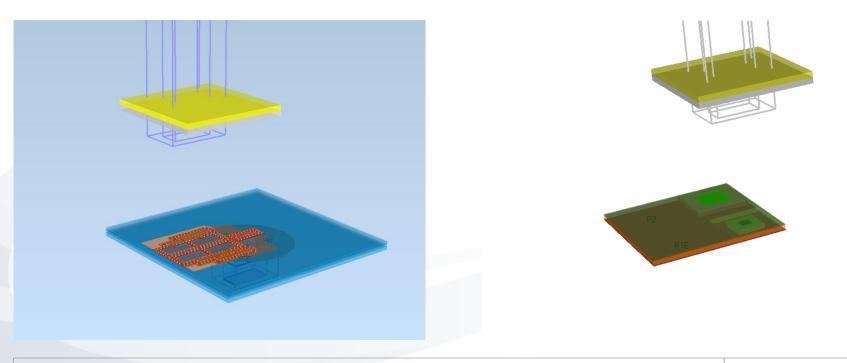




## > Investigation of Measurement Distance Tested Configurations

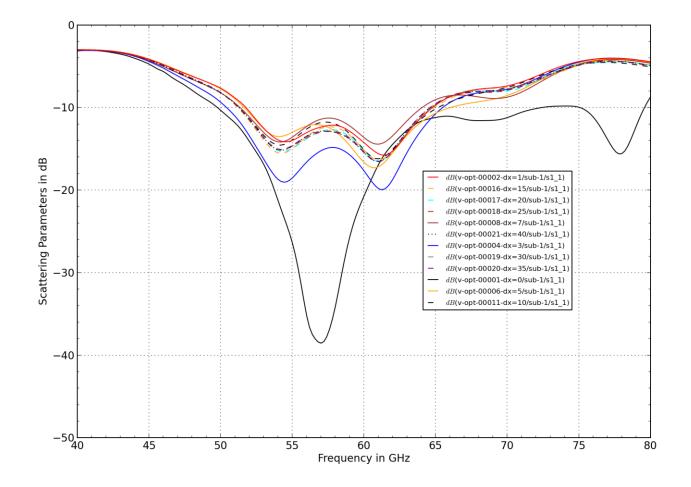


- $\rightarrow$  Simulated WG probe in the vicinity of the reference antennas
- $\rightarrow$  The WG probe is placed above the zone of maximum NF PD
- → The WG probe is gradually moved away from the DUT and the impact of the probe on the DUT is evaluated



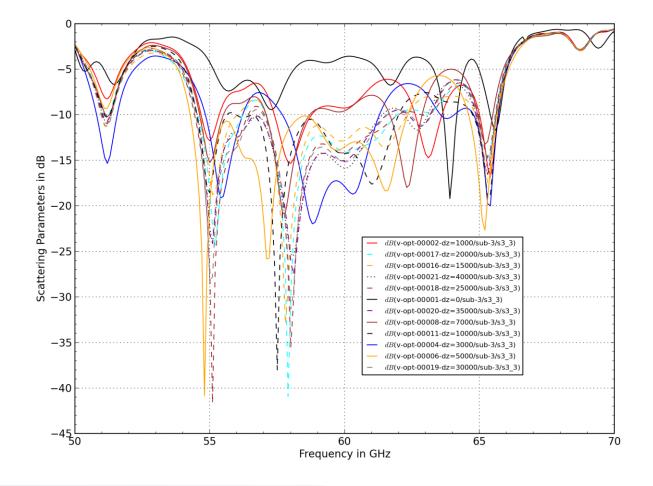


#### $\rightarrow$ Low impact on S11 with probe at >15 mm distance





#### $\rightarrow$ Low impact on S11 with probe at >25 mm distance







- → Two reference antenna simulation setups were made available as well as physical models
- → Accuracy of active near-field PD reconstruction using two-components of vector E-field has been demonstrated from numerical data
  - When measurement plane is at 25 mm from the DUT, deviation between simulated and reconstructed peak 1cm<sup>2</sup> spatial-averaged PD was found to be 13% at 2 mm from the DUT and 18% at 1 mm
  - H field reconstruction proved to be extremely accurate
  - Investigation on impact of measurement grid step shows that reconstruction accuracy does not degrade much with a step of up to  $\lambda/8$
- →  $5\lambda$  is a reasonable probing distance to avoid unacceptable perturbations of the DUT
- → A measurement scanner is being implemented to validate and further test near-field reconstruction algorithms: phased vs phaseless in practice?